

ferent parameters. The waveguide cross-section ratio is taken as $a/b=2$. The effect of the permittivity ϵ_r appears in Figs. 2 and 3, respectively, for slabs next to the waveguide wall and at the waveguide center, for $t/b=0.2$. Fig. 4 shows the effect of slab thickness for $\epsilon_r=9$ and $d_1=0$. The dotted lines show the appearance of the higher order modes.

To the readers who would be interested in other values of the parameters, for either *H*-plane or *E*-plane slabs, the authors would be pleased to send a copy of the computer listing, written in IBM 360 FORTRAN IV BOS.

When the dielectric slab is next to the guide wall (Figs. 2 and 4), the dominant mode is the LSM_{11} ("quasi TE_{01} "). The frequency bandwidth is limited by the appearance of the LSM_{21} mode at low values of t/b , or of the TE_{01} for large values of t/b .

When the slab is moved away from the guide wall (Fig. 3), the cutoff frequency of the TE_{01} mode decreases much faster than that of

the LSM_{11} mode when ϵ_r increases. Eventually, for large values of ϵ_r , the TE_{01} becomes the dominant mode, the bandwidth being then limited by the LSE_{11} mode. It is worthwhile noting that, for certain values of ϵ_r and t/b , the dispersion curves of the TE_{01} and LSM_{11} modes cross each other: at that frequency, the phase velocities of the two modes are the same, their group velocities being different. This effect may be of interest in the design of filters.

It should be noted that, in most practical devices making use of *H*-plane loading, the slabs do not extend entirely across the waveguide. The results obtained here are not directly applicable to such devices, but may nevertheless give valuable information. For instance, the gaps in the dielectric can be considered as perturbations of the complete *H*-plane structure, and approximate results can be obtained using perturbation techniques.

ACKNOWLEDGMENT

Thanks are extended to the personnel of the Louvain University Computation Center, where all the calculations were carried out.

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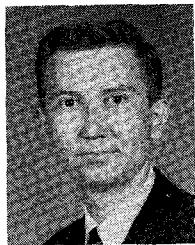
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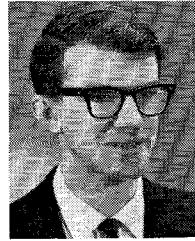


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